## Transform Coding

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11/24/2014

## Transform Domain Data Analysis

- Given an invertible transform $A$, the entropy of a source $\mathbf{x}$ does not change subject to $A$, i.e. $A \mathbf{x}$ has the same entropy as $\mathbf{x}$.
- However, there are several reasons why we want to perform lossy compression on $A \mathbf{x}$, instead of $\mathbf{x}$ :
- Input data sequence can be interpreted with more insights
- Input data possibly are de-correlated in transform domain
- The original time-ordered sequence of data can be decomposed into different categories


## Example: Height-Weight Data (1/3)

$\square$ The height-weight data pair tends to cluster alone the line $x_{h}=2.5 x_{w}$. A rotation transform

$$
A=\left(\begin{array}{cc}
\cos \phi & \sin \phi \\
-\sin \phi & \cos \phi
\end{array}\right), \quad \phi=68.02^{\circ},
$$

can simplify the data representation :



## Example: Height-Weight Data (2/3)

$\square$ If we set $\theta_{1}$ to zeros for all the data pairs, and transform the data back to $x_{h}-x_{w}$ domain, we have the reconstruction errors as follows:

| Original data |  |  |
| :---: | :---: | :---: |
| Height | Weight |  |
| 65 | 170 |  |
| 75 | 188 |  |
| 60 | 150 |  |
| 70 | 170 |  |
| 56 | 130 |  |
| 80 | 203 |  |
| 68 | 160 |  |
| 50 | 110 |  |
| 40 | 80 |  |
| 50 | 153 |  |
| 69 | 148 |  |
| 62 | 140 |  |
| 76 | 164 |  |
| 64 | -120 |  |

Reconstructed data

| Height | Weight |
| :---: | :---: |
| 68 | 169 |
| 75 | 188 |
| 60 | 150 |
| 68 | 171 |
| 53 | 131 |
| 81 | 203 |
| 65 | 162 |
| 45 | 112 |
| 34 | 84 |
| 60 | 150 |
| 61 | 151 |
| 57 | 142 |
| 67 | 168 |
| 50 | 125 |

## Example: Height-Weight Data (3/3)

$\square$ Note that, in original data, both $x_{h}$ and $x_{w}$ have nonnegligible variances, however, for $\theta_{0}$ and $\theta_{1}$, only $\theta_{0}$ has large variance
V Variance (or energy) of a source and its information has a positive relation; larger source variance, higher entropy

- For Gaussian source, the differential entropy is $\left(\log _{2} \pi e \sigma^{2}\right) / 2$.
- The error introduced into the reconstructed sequence of $\{x\}$ is equal to the error introduced into the transform-domain sequence $\{\theta\}$.


## Transform Coding Principle

- Transform step:
- The source $\left\{x_{n}\right\}$ is divided into blocks of size $N$. Each block is mapped into a transform sequence $\left\{e_{n}\right\}$ using a reversible mapping
- Most of the energy of the transformed block was contained in few elements of the transformed values
- Quantization step:
- The transformed sequence is quantized based on the following strategy:
- The desired average bit rate
- The statistics of the various transformed elements
- The effect of distortion on the reconstructed sequence
[ Entropy coding step:
- The quantized data are entropy-coded using Huffman, AC, or other techniques


## Transform Formulation

$\square$ For media coding, only linear transforms are used

- The forward transform can be denoted by

$$
\theta_{n}=\sum_{i=0}^{N-1} x_{i} a_{n, i} .
$$

- The inverse transform is

$$
x_{n}=\sum_{i=0}^{N-1} \theta_{i} b_{n, i} .
$$

- The selection of $N$ is application-specific
- Complexity of transform is lower for small $N$
- Large $N$ adapts to fast-changing statistics badly
- Large $N$ produces better resolution in transform domain


## 2-D Forward Transform

- For 2-D signals $X_{i, j}$, a general linear 2-D transform of block size $N \times N$ is given as

$$
\Theta_{k, l}=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} x_{i, j} a_{i, j, k, l} .
$$

If separable transform is used; the formulation can be simplified to

$$
\Theta_{k, l}=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} a_{k, i} x_{i, j} a_{j, l}=\sum_{i=0}^{N-1} a_{k, i}\left(\sum_{j=0}^{N-1} x_{i, j} a_{j, l}\right) \text {. }
$$

- In matrix form, the separable transform becomes

$$
\Theta=A \boldsymbol{X} A^{T} .
$$

## Orthonormal Transform

$\square$ All the transforms used in multimedia compression are orthonormal transforms. Thus, $A^{-1}=A^{T}$. In this case, $\Theta=A \boldsymbol{X} A^{T}$ becomes $\Theta=A \boldsymbol{X} A^{-1}$.

- Orthonormal transforms are energy preserving

$$
\begin{aligned}
\sum_{i=0}^{N-1} \theta_{i}^{2} & =\theta^{T} \theta=(\mathbf{A} \mathbf{x})^{T} \mathbf{A} \mathbf{x} \\
& =\mathbf{x}^{T} \mathbf{A}^{T} \mathbf{A} \mathbf{x}=\mathbf{x}^{T} \mathbf{x}=\sum_{n=0}^{N-1} x_{n}^{2}
\end{aligned}
$$

## Energy Compaction Effect

- The efficiency of a transform depends on how much energy compaction is provided by the transform
- The amount of energy compaction can be measured by the ratio of the arithmetic mean of the variances to their geometric means:

$$
G_{T C}=\frac{\frac{1}{N} \sum_{i=0}^{N-1} \sigma_{i}^{2}}{\left(\Pi_{i=0}^{N-1} \sigma_{i}^{2}\right)^{\frac{1}{N}}},
$$

where $\sigma_{i}^{2}$ is the variance of the $i$ th coefficients.

## Decomposition of 1-D Input

Transform decomposes an input sequence into components with different characteristics. If

$$
A=\frac{1}{\sqrt{2}}\left[\begin{array}{cc}
1 & 1 \\
1 & -1
\end{array}\right],
$$

input $\mathbf{x}=\left[x_{1}, x_{2}\right]$, the transformed output is

$$
A \mathbf{x}=\left[\frac{\left(x_{1}+x_{2}\right)}{\sqrt{2}}, \frac{\left(x_{1}-x_{2}\right)}{\sqrt{2}}\right] .
$$

The first transformed component computes the average (i.e. low-pass) behavior of the input sequence, while the $2^{\text {nd }}$ component captures the differential (i.e. high-pass) behavior of the input.

## Decomposition of 2-D Input

If $A$ in previous example is used for 2-D transform and $X$ is a 2-D input, we have $X=A^{T} \Theta A$ :

$$
\begin{aligned}
{\left[\begin{array}{ll}
x_{00} & x_{01} \\
x_{10} & x_{11}
\end{array}\right] } & =\frac{1}{2}\left[\begin{array}{cc}
1 & 1 \\
1 & -1
\end{array}\right]\left[\begin{array}{cc}
\theta_{00} & \theta_{01} \\
\theta_{10} & \theta_{11}
\end{array}\right]\left[\begin{array}{cc}
1 & 1 \\
1 & -1
\end{array}\right] \\
& =\frac{1}{2}\left[\begin{array}{ll}
\theta_{00}+\theta_{01}+\theta_{10}+\theta_{11} & \theta_{00}-\theta_{01}+\theta_{10}-\theta_{11} \\
\theta_{00}+\theta_{01}-\theta_{10}-\theta_{11} & \theta_{00}-\theta_{01}-\theta_{10}+\theta_{11}
\end{array}\right] \\
& =\theta_{00} \alpha_{0,0}+\theta_{01} \alpha_{0,1}+\theta_{10} \alpha_{1,0}+\theta_{11} \alpha_{1,1},
\end{aligned}
$$

where $\alpha_{i, j}$ is the outer product of $i$ th and $j$ th rows of $A$.
$\square$ How do you interpret $\theta_{0,0}, \ldots, \theta_{1,1}$ ?

- $\theta_{0,0}$ is the DC coefficient, and other $\theta_{i, j}$ are AC coefficients.


## Karhunen-Loeve Transform (KLT)

$\square$ KLT consists of the eigenvectors of the autocorrelation matrix: $[R]_{i, j}=E\left[X_{n} X_{n+1 i-j]}\right]$.
$\square$ KLT minimizes the geometric means of the variance of the transform coefficients $\rightarrow$ provides maximal $G_{T C}$

- Issues with KLT
- For non-stationary inputs, the autocorrelation function is time varying; computation of KLT is relatively expensive
- KLT matrix must be transmitted to the decoder
- If the input statistics change slowly, and the transform size can be kept small, the KLT can be useful


## Example: KLT

For $N=2$, the autocorrelation matrix for a stationary process is

$$
R=\left[\begin{array}{ll}
R_{x x}(0) & R_{x x}(1) \\
R_{x x}(1) & R_{x x}(0)
\end{array}\right],
$$

The eigenvectors of $R$ are

$$
v_{1}=\left[\begin{array}{l}
\alpha \\
\alpha
\end{array}\right], v_{2}=\left[\begin{array}{c}
\beta \\
-\beta
\end{array}\right] .
$$

With orthonormal constraint, the transform matrix is

$$
\mathbf{K}=\frac{1}{\sqrt{2}}\left[\begin{array}{cc}
1 & 1 \\
1 & -1
\end{array}\right] .
$$

## Discrete Cosine Transform

- DCT is derived from the Discrete Fourier Transform (DFT) by first perform an even-function extension to the input data, then compute its DFT:
- Only real number operations are required
- Better energy compaction than DFT


DCT


## DCT Formulation

- The rows of DCT matrix is composed of cosine functions of different frequencies:

$$
[C]_{i, j}=\left\{\begin{array}{lc}
\sqrt{\frac{1}{N}} \cos \frac{(2 i+1) j \pi}{2 N} & i=0, j=0,1, \ldots, N-1 \\
\sqrt{\frac{2}{N}} \cos \frac{(2 i+1) j \pi}{2 N} & i=1, \ldots, N-1, j=0,1, \ldots, N-1
\end{array} .\right.
$$

- The inner product of the input signal with each row of the matrix is the projection of the input signal onto a cosine function of fixed frequency
- The larger $N$ is, the better the frequency resolution is


## Basis Functions of 8-Point DCT

E Each column of the DCT matrix is a basis function:



## Basis Images of 8-Point 2-D DCT

- DCT can be extended to a 2-D transform:



## Performance of DCT

- For Markov sources with high correlation coefficient $\rho$,

$$
\rho=\frac{E\left[x_{n} x_{n+1}\right]}{E\left[x_{n}^{2}\right]},
$$

the compaction ability of DCT is close to that of KLT
$\square$ As many sources can be modeled as Markov sources with high values for $\rho$, DCT is the most popular transform for multimedia compression

## Discrete Walsh-Hadamard Trans.

- The Hadamard transform is defined by an $N \times N$ matrix $H$ with the property $H H^{T}=N I$.
- Simple to compute while still separate low frequency from high frequency components of the input data
- The Hadamard matrix is recursively defined as:

$$
H_{2 N}=\left[\begin{array}{cc}
H_{N} & H_{N} \\
H_{N} & -H_{N}
\end{array}\right] \text {, and } H_{1}=[1] .
$$

- The DWHT transform matrix is obtained by
- Normalize the matrix by $1 / N^{1 / 2}$ so that it is orthonormal
- Re-arrange the rows according to number of sign changes


## Coding of Transform Coefficients

- Different transform coefficients should be quantized and coded differently based on the amount of information it carries
- Information is related to the variance of each coefficients
$\square$ The bit allocation problem tries to determine the level of quantizer to use for different transform coefficients
- The Lagrange multiplier optimization technique is often used to solve the optimal bit allocation


## Lagrange Multiplier

$\square$ A constrained optimization problem tries to minimize a cost function $f(x, y)$ subject to some constraints on the parameter $x$ and $y: g(x, y)=c$

- The Lagrange cost function is defined as follows:

$$
J(x, y, \lambda)=f(x, y)-\lambda \cdot\|g(x, y)-c\|^{2} .
$$

$\square$ Solution: solve

$$
\nabla_{x, y, \lambda} J(x, y, \lambda)=0
$$



## Rate-Distortion Optimization (1/3)

If the rate per coefficient is $R$ and the rate per $k$ th coefficient is $R_{k}$, then

$$
R=\frac{1}{M} \sum_{k=1}^{M} R_{k},
$$

where $M$ is the number of transform coefficients
$\square$ The error variance for the $k$ th quantizer $\sigma_{r_{k}}^{2}$, is related to the $k$ th input variance $\sigma \theta_{k}^{2}$, by:

$$
\sigma_{r_{k}}^{2}=\alpha_{k} 2^{-2 R_{k}} \sigma_{\theta_{k}}^{2},
$$

where $\alpha_{k}$ depends on input distribution and quantizer

- The total reconstruction error is given by

$$
\sigma_{r}^{2}=\sum_{k=1}^{M} \alpha_{k} 2^{-2 R_{k}} \sigma_{\theta_{k}}^{2} .
$$

## Rate-Distortion Optimization (2/3)

$\square$ The objective of the bit allocation procedure is to find $R_{k}$ to minimize $\sigma_{r}^{2}$ subject to total rate constraint $R$.

- If we assume that $\alpha_{k}$ is a constant $\alpha$ for all $k$, we can set up the minimization problem in terms of Lagrange multipliers as

$$
J=\alpha \sum_{k=1}^{M} 2^{-2 R_{k}} \sigma_{\theta_{k}}^{2}-\lambda\left(R-\frac{1}{M} \sum_{k=1}^{M} R_{k}\right) .
$$

- Taking the derivative of $J$ with respect to $R_{k}$ and setting it to zero, we obtain the expression for $R_{k}$ :

$$
R_{k}=\frac{1}{2} \log _{2}\left(2 \alpha \ln 2 \sigma_{\theta_{k}}^{2}\right)-\frac{1}{2} \log _{2} \lambda
$$

## Rate-Distortion Optimization (3/3)

$\square$ Substituting $R_{k}$ to the expression for $R$, we have:

$$
\lambda=\prod_{k=1}^{M}\left(2 \alpha \ln 2 \sigma_{\theta_{k}}^{2}\right)^{\frac{1}{m}} 2^{-2 R} .
$$

- Therefore, the individual bit allocations for each transform coefficients is:

$$
R_{k}=R+\frac{1}{2} \log _{2} \frac{\sigma_{\theta_{k}}^{2}}{\prod_{k=1}^{M}\left(\sigma_{\theta_{k}}^{2}\right)^{\frac{1}{4}}} .
$$

- Note that $R_{k}$ may not be integers or positive numbers
- Negative $R_{k}$ 's are set to zero
- Positive $R_{k}$ 's are reduced to a smaller integer value


## Zonal Sampling

Z Zonal sampling is a simple bit allocation algorithm:

1. Compute $\sigma_{\theta_{k}}{ }^{2}$ for each coefficient.
2. Set $R_{k}=0$ for all $k$ and set $R_{b}=M R$, where $R_{b}$ is the total number of bits available for distribution.
3. Sort the variances $\left\{\sigma_{\theta_{k}}^{2}\right\}$ Suppose $\sigma_{\theta_{m}}{ }^{2}$ is the maximum.
4. Increment $R_{m}$ by 1 , and divide $\sigma_{\theta_{m}}{ }^{2}$ by 2.
5. Decrement $R_{b}$ by 1 . If $R_{b}=0$, then stop; otherwise, go to 3 .

| Bit allocation map for an $8 \times 8$ transform |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 | 7 | 5 | 3 | 1 | 1 | 0 | 0 |
| 7 | 5 | 3 | 2 | 1 | 0 | 0 | 0 |
| 4 | 3 | 2 | 1 | 1 | 0 | 0 | 0 |
| 3 | 3 | 2 | 1 | 1 | 0 | 0 | 0 |
| 2 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

## Threshold Coding

- Another bit allocation policy is called threshold coding
- Arrange the transform coefficients in a line
- The first coefficient is always coded
- For remaining coefficients
- If the magnitude is smaller than a threshold, it is skipped
- If the magnitude is larger than a threshold, its quantized value and the number of skipped coefficients before it is coded
$\square$ Zigzag scan is often used for 2-D to 1-D mapping



## JPEG Image Compression

- A standard defined by ISO/IEC JTC1/SC 29/WG 1 in 1992
- The official IS number is IS 10918-1, which defines the input to the decoder (a.k.a. the elementary stream), and how the decoder reconstructs the image
- The popular file format JFIF for JPEG elementary stream is defined in 10918-5
There are several new image coding standards that are incompatible to the old JPEG, but still bearing the JPEG name
- Wavelet-based JPEG-2000 (IS 15444-1)
- High quality lossless/lossy JPEG-XR (IS 29199-2)


## JPEG Initial Processing

$\square$ Color space RGB $\rightarrow Y C_{B} C_{R}$ mapping
$\square$ Chroma channel 4:2:2 sub-sampling

- Level shifting: assume each pixel has $p$-bit, then each pixel $x_{i, j}=x_{i, j}-2^{p-1}$
$\square$ Split pixels into $8 \times 8$ blocks
- If image size is not a multiple of 8 , extra rows/columns are padded to achieve multiple of 8
- Padded data is discarded after decoding


## JPEG 8×8 DCT Transform

$\square$ Forward DCT is applied to each $8 \times 8$ block


## JPEG Quantization

- Midtread quantization is used; the step size for each coefficients is from an $8 \times 8$ quantization matrix $Q$, e.g.,

| 16 | 11 | 10 | 16 | 24 | 40 | 51 | 61 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12 | 12 | 14 | 19 | 26 | 58 | 60 | 55 |
| 14 | 13 | 16 | 24 | 40 | 57 | 69 | 56 |
| 14 | 17 | 22 | 29 | 51 | 87 | 80 | 62 |
| 18 | 22 | 37 | 56 | 68 | 109 | 103 | 77 |
| 24 | 35 | 55 | 64 | 81 | 104 | 113 | 92 |
| 49 | 64 | 78 | 87 | 103 | 121 | 120 | 101 |
| 72 | 92 | 95 | 98 | 112 | 100 | 103 | 99 |

- Quantized values are called "labels." For input coefficient $\theta_{i j}$, we have

$$
l_{i j}=\left\lfloor\frac{\theta_{i j}}{Q_{i j}}+0.5\right\rfloor .
$$

## JPEG Quantization Example

- Quantization controls the entropy of the image
- Quantization matrices reflect image quality
- A scalar number (quality factor) is often used as quantization matrix multiplier to control image quality

$\theta_{00}$| 39.88 | 6.56 | -2.24 | 1.22 |
| ---: | ---: | ---: | ---: |
| -102.43 | 4.56 | 2.26 | 1.12 |
| 37.77 | 1.31 | 1.77 | 0.25 |
| -5.67 | 2.24 | -1.32 | -0.81 |


$Q_{00}$| 16 | 11 | 10 | 16 |
| :--- | :--- | :--- | :--- |
| 12 | 12 | 14 | 19 |
| 14 | 13 | 16 | 24 |
| 14 | 17 | 22 | 29 |

$$
l_{00}=\left\lfloor\theta_{00} / Q_{00}+0.5\right\rfloor=\lfloor 39.88 / 16+0.5\rfloor=\lfloor 2.99\rfloor=2
$$



## Entropy Coding

$\square$ DC/AC coefficients are coded differently

- DCs are coded using
- Differential coding + Huffman coding
- Each DC difference is coded using a Huffman prefix plus a fixed length suffix
- ACs are coded using
- Run-Length coding + Huffman coding


## DC Difference Code Table

| Difference <br> category <br> (VLC-code <br> as prefix) |  |  |  |  |  |
| :--- | ---: | ---: | ---: | ---: | ---: | ---: |
| 0 |  |  | 0 |  |  |

## AC RLE Code Table

$\square A C$ is zigzag scanned into a 1-D sequence
$\square$ Each non-zero coefficient is coded using a Z/C codeword plus a sign bit $S$
■ $Z$ - number of zero run before the label

- $C$ - label magnitude
- EOB is used to signal the end of each block
- ZRL is used to signal 15 consecutive zeros

| $Z / C$ | Codeword | $Z / C$ | Codeword | $\cdots$ | $Z / C$ | Codeword |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $0 / 0(\mathrm{EOB})$ | 1010 |  |  | $\cdots$ | $\mathrm{~F} / 0(\mathrm{ZRL})$ | 11111111001 |
| $0 / 1$ | 00 | $1 / 1$ | 1100 | $\cdots$ | $\mathrm{~F} / 1$ | 111111111110101 |
| $0 / 2$ | 01 | $1 / 2$ | 11011 | $\cdots$ | $\mathrm{~F} / 2$ | 1111111111110110 |
| $0 / 3$ | 100 | $1 / 3$ | 1111001 | $\cdots$ | $\mathrm{~F} / 3$ | 1111111111110111 |
| $0 / 4$ | 1011 | $1 / 4$ | 111110110 | $\cdots$ | $\mathrm{~F} / 4$ | 1111111111111000 |
| $0 / 5$ | 11010 | $1 / 5$ | 11111110110 | $\cdots$ | $\mathrm{~F} / 5$ | 1111111111111001 |

## JPEG Coding Example

- A good example from Wikipedia:


83,261 bytes
compression ratio 2.6:1


15,138 bytes
compression ratio 15:1


4,787 bytes compression ratio 46:1

